

Many-to-Many Communication for Mobile Ad Hoc Networks

Renato M. de Moraes, *Member, IEEE*, Hamid R. Sadjadpour, *Senior Member, IEEE*,
and J. J. Garcia-Luna-Aceves, *Fellow Member, IEEE*

Abstract—We introduce a collaboration-driven approach to the sharing of the available bandwidth in wireless ad hoc networks, which we call *many-to-many communication*, that allows concurrent multi-packet transmissions (MPTs) and multi-packet receptions (MPRs). Many-to-many communication also permits one-time multi-copy relaying of the same packet, which reduces the packet delivery delay compared to single-copy relaying without any penalty in capacity. Our scheme is based on the integration of multi-user detection and position-location information with frequency and code division in mobile ad hoc networks (MANETs). Transmissions are divided in frequency and codes according to node locations, and successive interference cancellation (SIC) is used at receivers to allow them to decode and use all transmissions from strong interfering sources. Consequently, the interference is divided into constructive interference (COI) and destructive interference (DEI). We show that, if each node is allowed to expand its bandwidth, both the link's Shannon capacity and the per source-destination throughput scale like $O(n^{\frac{\alpha}{2}})$ (upper-bound) and $\Omega[f(n)]$ (lower-bound), for n nodes in the network, a path loss parameter $\alpha > 2$, and $1 \leq f(n) < n^{\frac{\alpha}{2}}$.

Index Terms—Ad hoc network, FDMA/CDMA, Information Theory, mobility, MPR, MPT, SIC, throughput.

I. INTRODUCTION

THE protocol stacks of wireless ad hoc networks implemented or proposed to date have been designed to try to avoid interference. Hence, communication protocols used in wireless ad hoc networks today are meant to support reliable communication among senders and receivers that are *competing* with one another for the use of the shared bandwidth. This “competition-driven” view of bandwidth sharing has had profound implications on network architectures and methods used to access the channel and disseminate information. Gupta and Kumar [1] showed that, in a wireless connected network with static nodes, the throughput for each node degrades as the number of nodes increases under the competition-driven

view of networking. That is, it scales as $\Theta(1/\sqrt{n \log(n)})$,¹ where n is the number of nodes in the network.

Grossglauser and Tse [2] analyzed a two-hop, single-relay forwarding scheme for MANETs in which a source passes a packet to a relay that in turn delivers it to the destination when the two nodes are close to each other. This and many subsequent studies on how to make MANETs scale by using mobility [2], [3], [4], [5] consider each transmission as competing with all the other concurrent transmissions in the network. However, because a relay cooperates with a source by storing the source's packet until it is close enough to the intended destination, the throughput of MANETs can be increased.²

Recently, Toupmpis and Goldsmith [6] have shown that the capacity regions for ad hoc networks are increased when multiple access schemes are combined with spatial reuse (i.e., multiple simultaneous transmissions), multihop routing (i.e., packet relaying), and SIC, even without performing power control. Also, SIC circuits with simple implementation and low complexity have been introduced [7], and code division multiple access (CDMA) [8] and global positioning system (GPS) [9] technologies have been already integrated into a single IC chip [10]. Although CDMA and SIC for ad hoc networks have been studied in the past [11], [12], [13], [14], prior approaches have assumed that each transmission competes with others.

These works [1], [2], [3], [6], [11], [12], [13], [14], [5], characterize a one-to-one communication approach which stems from cellular concepts and in our opinion, it is not appropriate for ad hoc networks. Our earlier work [4] describes a setting for one-to-many communication. In this scenario, a node relays its packet to multiple relay nodes that are close, allowing them to cooperate to search for the destination. In this scheme, however, all the transmitting nodes in each communication session compete with each other to transmit their packets. Ghez et al. [15] and Tong et al. [16] explain a framework for many-to-one communication, i.e., multi-packet reception (MPR). In this context, multiple nodes cooperate to transmit their packets simultaneously to a single node using CDMA and the receiver node utilizes multiuser detection to decode multiple packets. Under this condition, two groups of multiple transmitting nodes that are close to each other have to compete with one another to transmit their packets to their respective receivers.

Manuscript received August 28, 2007; revised February 15, 2008 and June 3, 2008; accepted July 7, 2008. The associate editor coordinating the review of this paper and approving it for publication was P. Fan.

Part of this paper was presented and published at the IEEE Conference on Computer Communications (INFOCOM), Anchorage, AK, 2007.

R. M. de Moraes is with the Department of Computing and Systems, University of Pernambuco, Brazil (e-mail: renato@dsc.upe.br).

H. R. Sadjadpour is with the Department of Electrical Engineering, University of California at Santa Cruz (UCSC) (e-mail: hamid@soe.ucsc.edu).

J. J. Garcia-Luna-Aceves is with the Department of Computer Engineering at UCSC, and with Palo Alto Research Center, 3333 Coyote Hill Road, Palo Alto, CA 94304, USA (e-mail: jj@soe.ucsc.edu).

Digital Object Identifier 10.1109/TWC.2009.070961

¹ Ω , Θ and O are the standard order bounds. $\log(\cdot)$ is the natural logarithm.

² In [2], the per source-destination throughput scales as $\Theta(1)$.

send to and receive from many nodes (i.e., MPT and MPR in each node). Since full-duplex data communication in the same frequency band is not practical, we present an example of how many-to-many communication can be implemented with a scheme based on frequency division multiple access (FDMA) and CDMA that supports concurrent many-to-many transmissions.

A. Bandwidth Allocation and Data Packet Forwarding

In our specific implementation of many-to-many communication, we use two types of channels. Control channels are used by nodes to obtain such information as the identities of strong interference sources, the data packets expected by destinations, and the state of data channels (by virtue of training sequences). Nodes employ conventional digital transceivers for the control channels. Data channels are used to transmit data taking advantage of SIC at the receivers. Thus, there are two separate transmitter (receiver) circuits in each node [18]. One circuit is intended to transmit (receive) control packets, and the other is used to transmit (receive) data packets. Both circuits operate in different time and frequency with respect to each other.

Control (or Signaling) Channels: Due to cell arrangement (see Figs. 1 and 2), each cell is allocated a control frequency band from twelve⁷ non-overlapping control frequency bands, ω_1 to ω_{12} , to enable frequency reuse while avoiding interference in the control channels from nearby cells. Each control frequency band ω_i has a size of $|\omega_i| = \Delta\omega$ for $i = 1, \dots, 12$. Hence, the total bandwidth required for the control channels is $\Delta\omega_C = 12\Delta\omega$.

The maximum number of cells in a cluster associated to a given node is twelve. The number of cells and the cluster shape are chosen such that if the receiver range has maximum value, i.e., almost $\sqrt{2a(n)}$, then the receiver range reaches all these cells. Also, two cells employing the same control frequency band are kept at least $\sqrt{5a(n)}$ units away from each other, i.e., a safe guard-zone separation, thus guaranteeing asymptotic constant non-zero signal-to-noise and interference ratio (SNIR) as $n \rightarrow \infty$ [4] in the control channel, making signaling feasible and allowing control frequency reuse.

To determine which control channel a given node should use, each node is assumed to know its own position (but not the position of any other node) by utilizing a GPS circuit [9], [10], and to store a geographical map of the cells in the network with their associated control frequencies. The GPS is also used to provide an accurate common time reference to keep all nodes synchronized.

Each node uses the control channel receiver to listen to the control channel of the cell as well as to the other 11 control channels, in order to obtain the IDs and training sequences of the other nodes in its cell and in the cluster it perceives, while not transmitting during the neighbor discovery phase [18].

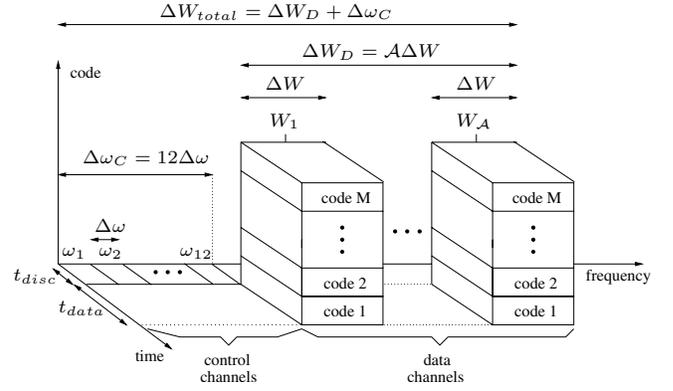


Fig. 2. Data and control channels spectra for the network.

Data Channels: To allow code reuse in the data channels while reducing the negative effects of interference, each cell is allocated a set of PN sequences (or codes) from the twelve different code sets available, ξ_1 to ξ_{12} , for communication in each data channel. Accordingly, each non-overlapping data channel is a half-duplex link of bandwidth ΔW . If \mathcal{A} ⁸ is the maximum number of nodes allowed to communicate in any cell, then $\Delta W_D = \mathcal{A}\Delta W$ is the data bandwidth required for the entire network and $M \geq 12\mathcal{A}$ distinct PN sequences are needed for local data communication. M is also called the spreading factor (or processing gain). Also, $\Delta W = B M$, where B is the original data bandwidth before spreading [8].

Because a PN sequence can be associated to a sequence of bits [8], they can be ordered and grouped as follows. $\xi_1 = \{C_1, \dots, C_{\mathcal{A}}\}$, $\xi_2 = \{C_{\mathcal{A}+1}, \dots, C_{2\mathcal{A}}\}$, ..., $\xi_{12} = \{C_{11\mathcal{A}+1}, \dots, C_{12\mathcal{A}}\}$, in which C_i stands for the i^{th} PN sequence (or code). In this way, any set of twelve cells, numbered from 1 to 12, has a different set of codes. Therefore, by construction, the cluster seen by any node is composed of cells having distinct numbers, and consequently, different codes.

As we discuss in Section III-B, the signaling in the control channel provides each node in a cell i knowledge of who the other nodes in this same cell are, and the node uses this information to choose a data channel to receive data, as well as to select a code for transmission from the available PN sequences in ξ_i based on its own and neighbor IDs, in the following order⁹: (i) The node with the highest ID in cell i is associated with the data channel ΔW centered at W_1 , as well as it is assigned the first PN sequence in ξ_i . (ii) The node with the second highest ID in cell i is associated with the data channel ΔW centered at W_2 , as well as it is assigned the second PN sequence in ξ_i , and this continues for all nodes in cell i . (iii) The data channels not utilized become idle in cell i . It happens in those cells where the number of nodes is less than \mathcal{A} .

In a communication session, each node only needs to know the nodes in its cell (obtained during the neighbor discovery phase) and the signal strengths received from them (by virtue of CDMA-SIC), in order to identify nodes in its receiver range. Note that nodes in a cell are very close to each other since

⁷Note that twelve is the minimum number which allows frequency reuse, but one can use a greater number.

⁸ \mathcal{A} is a network design parameter which depends on the decoding hardware capability [8].

⁹For clarity, we also indicate W_j as the data channel associated to node j .

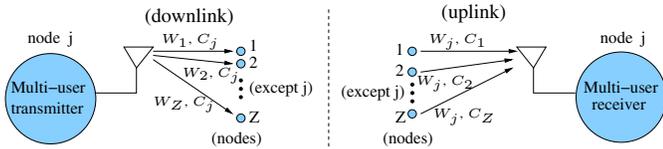


Fig. 3. Downlink and uplink description for data channels in a cell. Communication is Z -to- Z (i.e., many-to-many).

we are considering dense networks and near-far problem is not significant here. On those cases where near-far problem is an issue, frequency hopping spread spectrum may be employed to attain similar performance because communication among nodes are either direct or are relayed by other mobiles and there is no feasible method of power control to prevent near-far problem in sparse peer-to-peer communication networks [19].

With the deployment illustrated in Fig. 1, two or more nodes, while moving in the same cell, can perceive clusters composed of different cells with at most twelve distinct numbers. For example, in the middle of Fig. 1, node a , located exactly at the center of the cell 5, can apply SIC to decode the data signal from node b and node c in that same cell, each one being almost $\sqrt{a(n)}/2$ far apart from node a as shown (consequently, the receiver range for a is approximately $\sqrt{a(n)}/2$ and it is indicated by the dashed circle). Node a perceives the cluster composed of the five cells $\{2,4,5,6,9\}$ indicated in dashed line (i.e., those cells reached by a 's receiver range), and the other remaining closest four different cells $\{1,7,8,10\}$ are not necessary for decoding purposes. However, node b has to decode signals from nodes a and c which is almost $\sqrt{2a(n)}$ away (thus, the receiver range for b is approximately $\sqrt{2a(n)}$ and it is indicated by the solid circle). Hence, node b perceives the cluster with all the twelve cells $\{11,12,10,1,2,7,3,4,5,6,8,9\}$ shown in solid line, i.e., those cells reached by its receiver range. Analogously, node c perceives $\{2,7,4,5,6,11,8,9,10,1,12,3\}$ illustrated in dotted line. Therefore, by construction, the cluster perceived by any node is composed of cells having distinct numbers, and consequently, different codes.

At time t , each cell has Z nodes such that the data communication is Z -to- Z , i.e., many-to-many communication (see Fig. 3), where Z is a random variable due to mobility. Each node employs a multi-user CDMA transmitter [8] (i.e., it transmits up to $Z-1$ simultaneous data packets per session in which, due to FDMA, each packet is sent through a different data channel, as illustrated in Fig. 3(downlink)), spreading the data using the PN sequence associated to its ID. The node can transmit a different data packet in each channel or choose to send the same data packet in all (non-idle) channels, or a combination of both, depending on the fact that the node has packet for any destination in the same cell it is located. Thus, multi-copies of the same packet can be simultaneously relayed to reduce delay [4].

Given that each node is endowed with a multi-user detector (the SIC circuit) for its associated receiving data channel, it is able to decode the $Z-1$ simultaneous transmissions from all nodes in its cell (see Fig. 3(uplink)).

Data packet forwarding consists of two phases [2], [4]: The packet is transmitted from the source to possibly several

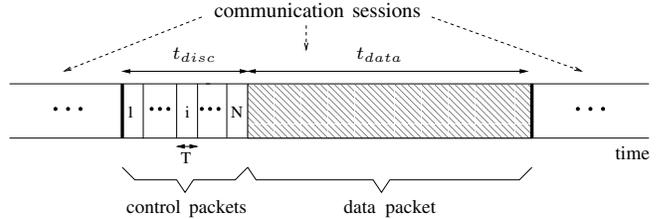


Fig. 4. Time series representation of control and data packets. t_{disc} is the neighbor discovery phase. t_{data} is the time period for transmission of data. t_{disc} plus t_{data} form a communication session.

relay nodes during *Phase 1* (i.e., one-time multi-copies can be forwarded), and it is delivered later to its destination by only one of the relay nodes during *Phase 2*. Both phases occur concurrently, but *Phase 2* has priority in all communications. These multiple one-time relays for the same packet provide better delay performance since the copies of the same packet follow different random routes, looking for the destination, reducing delay [4]. In the case of single-copy [2], the delay distribution behaves exponentially while in the case of multi-copy the random routes result in a delay distribution that has no tail, i.e., an exponential delay reduction is achieved with multi-copy forwarding [4].

B. Channel Access

Access to the channel is controlled by the signaling that takes place over the control channels. Such signaling occurs simultaneously in all cells, without suffering high interference from each other because of the different frequency assignment and consequent safe guard-zone separation (see Section III-A).

The signaling among the nodes in the same cell must be one-to-many and cannot assume knowledge of who the nodes in a cell are, because nodes are mobile. Each node needs to inform the other nodes in its present cell about its own presence in the cell, plus other control information. From Fig. 4, access to the channel is divided in time into a discovery phase and a data-transmission phase [20]. The period of "neighbor discovery" t_{disc} and the period for transmission of data t_{data} are constant and independent of the number of nodes in the network (n). Together, they compose a "communication session." The common time reference for communication sessions is obtained through the GPS circuit. The values of t_{disc} and t_{data} are system design parameters. t_{disc} is subdivided into N slots, each of length T . Hence, $T = \frac{t_{disc}}{N}$, where N is a positive integer to calculate according to some given criterion as explained later.

Each node simultaneously senses the channel to detect collision while transmitting in the control channel (e.g., using echo cancelling techniques [21], [22]). Accordingly, the nodes involved in a collision do not participate in that session anymore, i.e., they remain silent until the next session. Also, since only \mathcal{A} codes are available per cell, then, only the first \mathcal{A} nodes that successfully announced their control packets during t_{disc} are going to transmit (or receive) data during t_{data} for that session. Each control packet conveys, as a minimum, the node ID, a short training sequence and the sequence number (SN) the node is expecting as destination, while a data packet bears long sequences of bits. Therefore, we assume that $t_{disc} \ll t_{data}$.

TABLE I
PERFORMANCE VALUES FOR THE NEIGHBOR DISCOVERY PHASE FOR
 $\phi = 1/3$. \mathbb{P}_c REFERS TO Z_{max} FROM LEMMA 1.

n	Z_{max}	N_{min}	\mathbb{P}_c
1000	25	77	0.04
10^6	28	134	0.02
10^9	33	210	0.01

Each time the discovery period is about to begin, each node randomly chooses one of the N mini-slots and transmits its control packet. If there is no collision, i.e., if the other nodes in the same cell choose different mini-slots to transmit, then all the other nodes in the cell will receive this packet.¹⁰ A collision happens every time two or more nodes in the same cell choose to transmit in the same mini-slot. Let Z_i be the number of nodes in the same cell choosing the mini-slot i to transmit their control packets. Let Z_{max} be the maximum number of nodes in any cell. The probability of collision in mini-slot i , $\mathbb{P}_c(i)$, is given by

$$\mathbb{P}_c(i) = \mathbb{P}\{Z_i \geq 2\} = 1 - \left(1 - \frac{1}{N}\right)^{Z_{max}} - \frac{Z_{max}}{N} \left(1 - \frac{1}{N}\right)^{Z_{max}-1}. \quad (1)$$

The criterion used to choose N is as follows. We calculate N such that there is no collision in mini-slot i with probability approaching 1 as $n \rightarrow \infty$, for example, with probability $\geq 1 - \frac{\log(\log(n))}{\log(n)}$. From (1), $\mathbb{P}_c(i) \leq 1 - \left(1 - \frac{1}{N}\right)^{Z_{max}}$. Accordingly, we choose

$$\mathbb{P}_c(i) \leq 1 - \left(1 - \frac{1}{N}\right)^{Z_{max}} \leq \frac{\log(\log(n))}{\log(n)} \\ \implies N \geq \left\lceil \left[1 - \left(1 - \frac{\log(\log(n))}{\log(n)}\right)^{Z_{max}} \right]^{-1} \right\rceil = N_{min}, \quad (2)$$

in which $\lceil x \rceil$ stands for the ceil function (i.e., the smallest integer greater than or equal to x), and N_{min} is the actual value to be implemented for N . Thus, we have

$$T = \frac{t_{disc}}{N_{min}}. \quad (3)$$

The relationship between Z_{max} and n is given by the following lemma, which proof can be found in [24].

Lemma 1 *For the uniform mobility model, with probability approaching 1 as $n \rightarrow \infty$, the maximum number of nodes in any cell is given by*

$$Z_{max} = \left\lceil \frac{3 \log(n)}{\log(\log(n^\phi))} \right\rceil. \quad (4)$$

Fig. 5 illustrates the behavior of Z_{max} as a function of n for $\phi = 1/3$ from Lemma 1 and from simulations for the uniform mobility model [3] performed in MATLAB [25], which shows that the simulated maximum number of nodes in a cell is lower than that indicated by the lemma; however, as the total number of nodes n increases the simulation result approximates the predicted value by the lemma which becomes precise as $n \rightarrow \infty$. Table I shows the values attained by Z_{max} , N_{min} , and \mathbb{P}_c , for different values of n for $\phi = 1/3$. As the table shows, the probability of collision \mathbb{P}_c remains very low for a

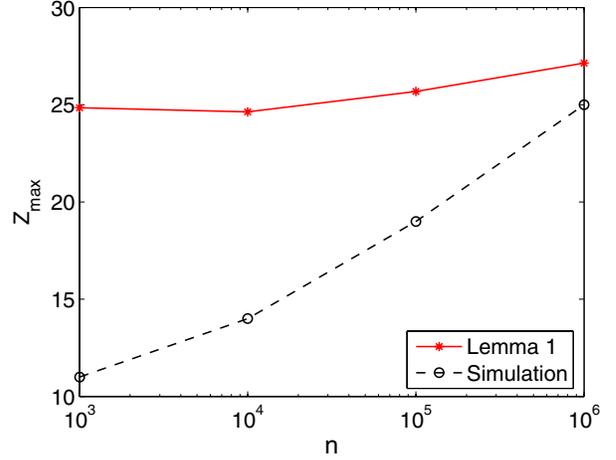


Fig. 5. Maximum number of nodes Z_{max} in a cell as a function of total number of nodes n in the network for $\phi = 1/3$ from Lemma 1 and from simulations of the uniform mobility model [3].

wide range of values of n . Also note that having a lower value for Z_{max} for finite n , as indicated by simulations, implies that the actual probabilities of collisions are lower than that showed in Table I which is good for general network performance. In addition, observe that collisions are even rarer within cells having fewer nodes than Z_{max} . Moreover, we defined the criterion for collision such that $\mathbb{P}_c \rightarrow 0$ as $n \rightarrow \infty$.

Although Z_{max} is the maximum number of nodes in any cell, in practice, the number of codes to be used is limited. Thus, at most \mathcal{A} nodes in any cell are allowed to get a code and communicate during t_{data} . However, Z_{max} grows very slowly with n . Thus, by choosing, for example, $\mathcal{A} \geq 10$, for practical values of ϕ , the fraction of cells having more than \mathcal{A} nodes can be bounded by a small constant, for n large. Accordingly, the total number of cells in the network ($\#$ of cells) = $1/a(n) = \phi n$. By considering the uniform mobility model, the fraction of cells containing $Z = j$ nodes is obtained by

$$\mathbb{P}\{Z = j\} = \binom{n}{j} \left(\frac{1}{\phi n}\right)^j \left(1 - \frac{1}{\phi n}\right)^{n-j} \xrightarrow{n \rightarrow \infty} \frac{1}{j!} \left(\frac{1}{\phi}\right)^j e^{-1/\phi}. \quad (5)$$

The fraction of cells having more than \mathcal{A} nodes as $n \rightarrow \infty$, for given ϕ , can be upper-bounded by

$$\mathbb{P}\{Z > \mathcal{A}\} = \sum_{j=\mathcal{A}+1}^{\infty} \frac{1}{j!} \left(\frac{1}{\phi}\right)^j e^{-1/\phi} \leq \frac{1}{(\mathcal{A}+1)!} \left(\frac{1}{\phi}\right)^{\mathcal{A}+1}. \quad (6)$$

For example, for $\phi = \frac{1}{3}$ and $\mathcal{A} = 10$, $\mathbb{P}\{Z > \mathcal{A}\} \leq 0.0044$ as $n \rightarrow \infty$.

C. Interference in a Data Channel

The interference in the data channel at a node j , regarding node i transmitting to node j through W_j , is defined as the signals coming from all transmitting nodes in the network, via W_j , except node i . It can be decomposed in the following two types.

Destructive Interference (DEI) for the node j comes from nodes, transmitting in W_j , outside the receiver range of j . *DEI* constitutes the part of the interference that will not be decoded.

¹⁰Note that this access method is a simple variation of standard access schemes [23].

$$SNIR = \frac{P_{i_j}(t)g_{i_j}(t)}{BN_0 + \underbrace{\frac{1}{M} \sum_{\substack{k \in range \\ k \neq i}} P_{k_j}(t)g_{k_j}(t)}_{COI} + \underbrace{\frac{1}{M} \sum_{\substack{k \notin range \\ C_k \neq C_i}} P_{k_j}(t)g_{k_j}(t) + \sum_{\substack{k \notin range \\ C_k = C_i}} P_{k_j}(t)g_{k_j}(t)}_{DEI}}. \quad (7)$$

Constructive Interference (COI) comes from nodes, transmitting in W_j , within the receiver range of j . By construction (see Section III-A), the nodes within the receiver range of j , transmitting in W_j , use different codes. *COI* constitutes the decodable part of the interference. The SNIR can be written as shown in (7).

If node i transmits data to j at time t , via W_j , the SNIR at the receiver j , without SIC, is given by (7) [2], where *range*¹¹ is the set of nodes transmitting in W_j and reached by the receiver range of node j , C_i is the PN sequence used by sender node i , $P_{i_j}(t) = P \forall(i, j)$ is the transmit power chosen by node i to transmit to node j (i.e., $P_{i_j}(t)$ is constant for all pair (i, j)), $g_{i_j}(t)$ is the channel path gain from node i to j , B is the original bandwidth of the data signal (before spreading), BN_0 is the noise power (where N_0 is the noise power spectral density), M is the spreading factor, *COI* and *DEI* are the total interference in W_j at node j . The summation terms in the denominator of (7) containing the factor $1/M$ constitute the multiple access interference (*MAI*). *MAI* is caused by partial cross-correlation among the distinct codes due to the asynchronous nature of the uplink channel [8]. The last summation term (without the factor $1/M$) is consequence of code reuse in the network and we call it *same code interference (SCI)*. Thus, $SCI = \sum_{\substack{k \notin range \\ C_k = C_i}} P_{k_j}(t)g_{k_j}(t)$, such that, $MAI + SCI = COI + DEI$. *MAI* and *SCI* presentations are easier for calculating SNIR as explained later.

The channel path gain g_{i_j} is assumed to be a function of the distance only (i.e., the simple path propagation model) [1], [2], therefore, $g_{i_j}(t) = \frac{1}{|X_i(t) - X_j(t)|^\alpha} = \frac{1}{r_{i_j}^\alpha(t)}$, in which α is the path loss parameter, and $r_{i_j}(t)$ is the distance between i and j .

D. Hybrid FDMA/CDMA Data Transceiver

From Fig. 3(downlink), the FDMA/CDMA data transmitter in node j selects packets previously relayed to node j which have their destination nodes present in the same cell, spread the data using the code C_j assigned to node j , and transmits each one of them through each different frequencies associated to each distinct destination node. If the node assigned to a data channel is not a destination for a relayed packet, then the transmitter selects a new packet generated locally by node j . Details on the data transceiver circuit can be found on [18].

The basic decoding scheme of the CDMA-SIC data receiver scheme is given in [7] (see also Fig. 3(uplink)), in which the decoding is performed successively from the strongest signal to the weakest. The use of training sequences obtained through the control channels allow to obtain a local estimation of the wireless channel. Thus, with the simple path propagation model assumed, the strongest signal decoded first comes from

the closest neighbor to node j (not necessarily in the same cell of j but in the cluster it perceives), while the weakest signal of interest (decoded last) is the farthest node to node j in the cell node j is located. Let *MAI'* be the remaining multiple access interference at node j after applying SIC up to node i , i.e.,

$$MAI' = \frac{1}{M} \sum_{\substack{k: g_{k_j} < g_{i_j} \\ C_k \neq C_i}} P_{k_j}(t)g_{k_j}(t). \quad (8)$$

Therefore, the resulting SNIR (called *SNIR'*) from node i to node j after applying SIC is given by

$$SNIR' = \frac{P_{i_j}(t)g_{i_j}(t)}{BN_0 + MAI' + SCI}. \quad (9)$$

Note that, depending on the position of the node j , it may have nodes transmitting from adjacent cells closer than a far node in the same cell. Therefore, j has to be able to decode the data signals from these adjacent cell nodes before decoding the signal from the far node of the same cell. This explains why each node also needs to obtain the training sequences from the other nodes located outside its cell but still within its receiver range. The receiver uses the information obtained during the neighbor discovery phase to retain the data packets from nodes in the same cell as j , dropping the outside cell packets since node j cannot keep track of all nodes in adjacent cells to see if this packet is for relaying or destination. Besides, from (7), SIC is fundamental to derive (9) and a node have all packets from the same cell successfully decoded.

IV. CAPACITY AND BANDWIDTH ANALYSIS

A. Link's Shannon Capacity

The link's Shannon capacity R_{i_j} in the data channel W_j , in which node j receives from node i , after j applying SIC up to node i , from (9), is given (in units of nats) by [26]

$$R_{i_j} = B \log(1 + SNIR') = B \log \left[1 + \frac{P}{BN_0 + MAI' + SCI} \right]. \quad (10)$$

The calculation considers the transmitting node i located at a distance $c_1 \sqrt{a(n)}$ from j , while due to SIC, all the remaining interfering nodes are at a distance greater than $c_1 \sqrt{a(n)}$ from j , where $c_1 \in (0, \sqrt{2})$ depends on the distance between node j and all other nodes in the cell. *MAI'* can be computed by using Fig. 6. Assume that the center of the unit square area is the origin O of the (x, y) coordinates, and that, at time t , the receiver node j is located at the point Q with coordinates $(x_Q, y_Q) \in (-\frac{1}{2}, \frac{1}{2})$. We divide the square unit area network in four triangles (indicated by MAI'_l for $l = 1, 2, 3, 4$ in Fig. 6) and compute the interference generated from each of these regions, such that $MAI' = \sum_{l=1}^4 MAI'_l$. Similar to [4], for a uniform distribution of the nodes, we

¹¹ $k \notin range$ means the nodes outside the receiver range of node j transmitting in W_j .

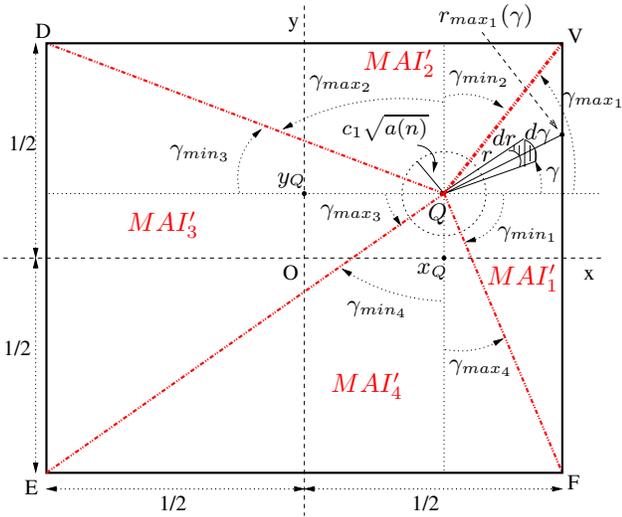


Fig. 6. Interference regions for node i communicating with node j . The angle γ increases in the counterclockwise direction.

consider a differential element area $r dr d\gamma$ that is distant r units from node j . Since the nodes are uniformly distributed and n grows to infinity, the node density in the network is $\frac{n}{1}$, and the summation in (8) can be bounded by an integral. Thus, MAI'_l at node j is upper-bounded by

$$MAI'_l(n) \leq \int_{region\ MAI'_l} \int \frac{P}{Mr^\alpha} \phi \epsilon_j \frac{n}{1} r dr d\gamma, \quad (11)$$

in which ϵ_j is the fraction of cells using the bandwidth W_j . Accordingly, ϵ_j equals the fraction of cells containing at least j nodes, in which $j \in [2, \mathcal{A}]$. From (5), we have as $n \rightarrow \infty$

$$\epsilon_j = \mathbb{P}\{Z \geq j\} = 1 - \sum_{k=0}^{j-1} \frac{1}{k!} \left(\frac{1}{\phi}\right)^k e^{-1/\phi}. \quad (12)$$

Thus, for $\alpha > 2$, and using that $a(n) = \frac{1}{\phi n}$, from (11) we obtain with some manipulations

$$\begin{aligned} MAI'_l(n) &\leq \int_{\gamma_{min_l}}^{\gamma_{max_l}} \int_{c_1 \sqrt{a(n)}}^{r_{max_l}(\gamma)} \frac{P \phi \epsilon_j n}{Mr^{\alpha-1}} dr d\gamma \\ &\leq c_2 \epsilon_j n^{\frac{\alpha}{2}} \left(1 - \frac{c_3}{n^{\frac{\alpha}{2}-1}}\right), \end{aligned} \quad (13)$$

in which c_2 and c_3 are positive constants for given l , (x_Q, y_Q) , c_1 , ϕ , M , P , and α . Therefore,

$$MAI' = \sum_{l=1}^4 MAI'_l \leq c_4 \epsilon_j n^{\frac{\alpha}{2}} \leq c_4 n^{\frac{\alpha}{2}}, \quad (14)$$

since $\epsilon_j \in [0, 1]$, and $\left(1 - \frac{c_3}{n^{\frac{\alpha}{2}-1}}\right) \leq 1$ for n large. c_4 is a positive constant function of the location (x_Q, y_Q) of node j .

On the other hand, the same code interference (SCI) can be upper-bounded by using the same procedure as done before for MAI' . Consequently, it can be shown that

$$SCI = \sum_{\substack{k \notin range \\ C_k = C_i}} P_{k_j}(t) g_{k_j}(t) \leq c_5 \epsilon_j n^{\frac{\alpha}{2}} \leq c_5 n^{\frac{\alpha}{2}}. \quad (15)$$

Hence, from (14) and (15), it results that the total remaining interference after SIC at node j is upper-bounded by

$$MAI' + SCI \leq (c_4 + c_5) n^{\frac{\alpha}{2}}. \quad (16)$$

Therefore, if we consider the expansion $B = f(n)$ of the original data bandwidth,¹² such that $1 \leq f(n) < n^{\frac{\alpha}{2}}$, then, a

¹²In our analysis, the bandwidth expansion is used to obtain the asymptotic behavior of the capacity.

lower-bound for R_{ij} can be obtained by using the maximum interference. Thus, from (10) and (16), the corresponding link's Shannon capacity lower-bound as $n \rightarrow \infty$, for node j receiving from node i , is obtained by

$$R_{ij} \geq f(n) \log \left(1 + \underbrace{\frac{c_6 n^{\frac{\alpha}{2}}}{f(n) N_o + (c_4 + c_5) n^{\frac{\alpha}{2}}}}_{n \rightarrow \infty c_7} \right) = c_7 f(n), \quad (17)$$

in which c_6 and c_7 are positive constants for given α , ϕ , P , M , N_o , c_1 , and (x_Q, y_Q) . In (17), *interference dominates noise for the bandwidth expansion* $1 \leq B < n^{\frac{\alpha}{2}}$.

On the other hand, if we consider a scenario such that there is no limitation on available bandwidth, then we can obtain an upper-bound for R_{ij} . Accordingly, from (10),

$$R_{ij} = B \log \left(1 + \frac{c_6}{\frac{BN_o}{n^{\frac{\alpha}{2}}} + \frac{1}{n^{\frac{\alpha}{2}}}(MAI' + SCI)} \right). \quad (18)$$

Now, from (16) and (18), and by taking $B \geq c_8 n^{\frac{\alpha}{2}}$, for some positive constant c_8 and n sufficiently large, it results that

$$\frac{1}{n^{\frac{\alpha}{2}}} (MAI' + SCI) \leq c_4 + c_5 \leq \frac{BN_o}{n^{\frac{\alpha}{2}}}. \quad (19)$$

Thus, the term $\frac{BN_o}{n^{\frac{\alpha}{2}}}$ becomes dominant in the denominator of (18) when $B \geq c_8 n^{\frac{\alpha}{2}}$ and $n \rightarrow \infty$. From (18) and (19), for $B \geq c_8 n^{\frac{\alpha}{2}}$, we have the following upper-bound for the link's Shannon capacity as $n \rightarrow \infty$

$$R_{ij} = n^{\frac{\alpha}{2}} \underbrace{\frac{B}{n^{\frac{\alpha}{2}}}}_{n \rightarrow \infty c_9} \log \left(1 + \frac{c_6}{\frac{BN_o}{n^{\frac{\alpha}{2}}} + \frac{1}{n^{\frac{\alpha}{2}}}(MAI' + SCI)} \right) = c_9 n^{\frac{\alpha}{2}}, \quad (20)$$

in which c_9 is a positive constant. *Here, noise dominates interference due to the large bandwidth expansion.*

Thus, (17) and (20) describe two limiting cases. The former is the minimum capacity attained if we use the bandwidth expansion $1 \leq B < n^{\frac{\alpha}{2}}$. The latter is the maximum capacity reachable if the available bandwidth is large such that $B \geq c_8 n^{\frac{\alpha}{2}}$. Note that any increase in B beyond $c_8 n^{\frac{\alpha}{2}}$ will not change the order of the upper-bound of the capacity.

B. Per Source-Destination Throughput

From Section III-B, each node accesses the data channel at a constant rate $\delta = \frac{t_{data}}{t_{disc} + t_{data}}$ with probability approaching 1 as $n \rightarrow \infty$, such that each source sends one packet per session to its destination. Each node is guaranteed, in each data channel, a communication rate of R_{ij} lower- and upper-bounded by (17) and (20), respectively. Also, this available communication rate has to be divided among all routes the node must serve per session per channel. However, due to the mobility and the routing scheme, each node serves only one route per session per data channel, i.e., the node either relays a new packet or it delivers a packet to a destination. Thus, the number of routes every node has to service per session per data channel is (# of served routes) = 1. Moreover, all cells containing at least two nodes are able to execute FDMA/CDMA and SIC successfully. From (5), $\mathbb{P}\{Z \geq 2\} = \left(1 - e^{-1/\phi} - \frac{1}{\phi} e^{-1/\phi}\right)$, as $n \rightarrow \infty$. Hence, with probability approaching 1 as $n \rightarrow \infty$,

the per source-destination throughput $\lambda(n)$ is obtained by [5], [4]

$$\lambda(n) = \frac{R_{ij} \delta \mathbb{P}\{Z \geq 2\}}{\# \text{ of served routes}} = c_{10} R_{ij}, \quad (21)$$

where c_{10} is a positive constant for given t_{disc} , t_{data} , and ϕ . From (17), (20), and (21), we proved the following theorem.

Theorem 1 *By employing mobility, CDMA, SIC, one-time relaying of packets, and bandwidth expansion using the many-to-many communication strategy, the ad hoc network attains, with probability approaching 1 as $n \rightarrow \infty$, the upper- and lower-bound per source-destination throughput given respectively by*

$$\lambda(n) = O(n^{\frac{\alpha}{2}}) \text{ and } \lambda(n) = \Omega[f(n)], \quad (22)$$

where $1 \leq f(n) < n^{\frac{\alpha}{2}}$.

C. Bandwidth Scalability

The total bandwidth requirement (ΔW_{total}) for the entire network has two components. One from the control channels ($\Delta \omega_C$), and the other from the data channels (ΔW_D).

From (2) and (3), Lemma 1, and noting that $\Delta \omega$ in each control channel equals $2/T$, due to the Nyquist rate, it results that

$$\Delta \omega_C = \frac{24N_{min}}{t_{disc}} = \Theta \left[1 - \left(1 - \frac{\log(\log(n))}{\log(n)} \right)^{\frac{1}{\left| \frac{3 \log(n)}{\log(\log(n^{\phi}))} \right|}} \right]^{-1}. \quad (23)$$

From Section III-A, $\Delta W = BM = 12AB$ [8]. Thus, the bandwidth scalability in each data channel associated to the upper- and lower-bound capacity is given respectively by

$$\Delta W = \Omega(n^{\frac{\alpha}{2}}) \text{ and } \Delta W = \Theta[f(n)], \quad (24)$$

where $1 \leq f(n) < n^{\frac{\alpha}{2}}$.

The total bandwidth for the entire network is obtained by

$$\Delta W_{total} = \Delta W_D + \Delta \omega_C = \mathcal{A} \Delta W + \Delta \omega_C, \quad (25)$$

where ΔW and $\Delta \omega_C$ are given above.

V. COMPARISON WITH PREVIOUS SCHEMES

A. The Static Network Case

Here, we consider the capacity and bandwidth expansion performance of the CDMA-SIC MAC layer scheme assuming a static network, and compare it with the results by Negi and Rajeswaran [12], which to the best of our knowledge is the only work that employed bandwidth expansion in the model presented by Gupta and Kumar [1].

The model we consider is that of a network formed by n fixed nodes, independently and uniformly distributed on a unit sphere surface. This model is also known as random network [1]. The set of assumptions assumed here are basically the same as those adopted in the previous sections, except that the n nodes are considered to be static on a unit sphere surface, and that the routing of packets is done through multiple hops along cells following the minimum distance path from source to destination. These assumptions are also consistent with the works by Negi and Rajeswaran [12] and Gupta and Kumar [1]. Because the communication framework is one-to-one, we use only one data channel ΔW for the entire network.

Therefore, no simultaneous data channels are needed given that only one copy of each packet is relayed along the route to its destination, and so we can apply the CDMA-SIC without FDMA.

In this model, the surface of the sphere is divided into cells. The number of nodes in any cell is a random variable Z . A node is randomly chosen to relay all traffic in each cell, and is called the head node of the cell. Thus, to guarantee relaying of traffic between cells, it is required that every cell has at least one node *whp* [1], i.e., $\mathbb{P}\{Z \geq 1\} \xrightarrow{n \rightarrow \infty} 1$. Within a cell, all sources send traffic to the head node, and destinations receive traffic from the head node.

Gupta and Kumar [1] showed that there exists a Voronoi tessellation \mathcal{V}_n on the unit sphere surface satisfying the following properties:

- Every Voronoi cell V contains a disk of area $100 \log(n)/n$ and corresponding radius $\rho(n) = c_{11} \sqrt{\log(n)/n}$.
- Every Voronoi cell is contained within a circle of radius $2\rho(n)$.

Each Voronoi cell $V \in \mathcal{V}_n$ is simply a cell of the network, and the cells do not have a regular shape because the network is random. With this tessellation, each cell contains at least one node *whp* which meets the connectivity requirement [1]. Furthermore, by choosing the transmission range equal to $8\rho(n)$ for each node, it allows direct communication within a cell and between adjacent cells. Accordingly, two cells are interfering neighbors if there is a point in one cell that is within a distance $(2 + \Delta)8\rho(n)$ of some point in the other cell, in which $\Delta > 0$ is a given constant modeling situations where a guard zone is required to prevent a neighboring node from transmitting on the same channel at the same time [1].

Another useful property of this Voronoi tessellation is that every cell $V \in \mathcal{V}_n$ has no more than c_{12} interfering neighbors, and hence the maximum number of interfering nodes is bounded by some positive constant [1]. Consequently, similarly to what we did in the mobile case, we can assign distinct PN sequences to each node, such that every cell in \mathcal{V}_n has interfering neighbors using different codes. Therefore, we need $M \geq c_{12}$ distinct PN sequences and reuse the codes in order to save bandwidth. Note that GPS (or some other technique) is no longer required since nodes are static. However, as explained before, MAI has to be considered even when transmission synchronization among nodes is employed [8]. Because nodes are static, we only need to assign the different codes during the initialization of the network.

We compute the link's Shannon capacity for an arbitrary pair of nodes from adjacent cells, noting that the analysis applied for the mobile network can be used for the static network as well. Thus, similarly to the description in Section III-C, each node communicating with another node applies SIC to eliminate MAI from close neighbors and the $SNIR'$ computation follows (9). The MAI calculation is done following an approach similar to that of (13), but considering the unit sphere surface. Furthermore, because the communication is either between two nodes on the same cell or between two head nodes from adjacent cells, any two communicating nodes are located at distance $c_{13}\rho(n)$ apart from each other. Therefore, if node j is receiving data from node i , after node

j applies SIC up to node i , all the remaining interfering nodes are placed at distance greater than $c_{13}\rho(n)$. Hence, for $\alpha > 2$ and n sufficiently large, we have the following bound for the remaining multiple access interference at a node j receiving data from node i , after SIC,

$$\begin{aligned} MAI'(n) &\leq \int_0^{2\pi} \int_{c_{13}\rho(n)}^{\frac{\sqrt{\pi}}{2}} \frac{P\phi n}{Mr^{\alpha-1}} dr d\gamma \\ &= \frac{2\pi P\phi n}{M(\alpha-2)} \left[\left(\frac{n}{c_{14} \log(n)} \right)^{\frac{\alpha}{2}-1} - \left(\frac{2}{\sqrt{\pi}} \right)^{\alpha-2} \right] \\ &\leq \frac{c_{15} n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}}, \end{aligned} \quad (26)$$

where c_{15} is a positive constant given that c_{11} , c_{13} , ϕ , M , P , and α are specified.

Analogously, the same code interference (*SCI*) can be upper-bounded by

$$SCI = \sum_{\substack{k \notin \text{range} \\ C_k = C_i}} P_{kj}(t) g_{kj}(t) \leq \frac{c_{16} n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}}. \quad (27)$$

Hence, from (26) and (27), the total remaining interference after SIC at node j is upper-bounded by

$$MAI' + SCI \leq (c_{15} + c_{16}) \frac{n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}}. \quad (28)$$

From (10), assuming that node i transmits to j , in which the original data bandwidth is B (before spreading), we obtain the following link's Shannon capacity

$$\begin{aligned} R_{ij} &= B \log \left(1 + \frac{P}{BN_o + MAI' + SCI} \right) \\ &= B \log \left(1 + \frac{c_{17}}{\frac{BN_o (\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}} + \frac{(\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}} (MAI' + SCI)} \right). \end{aligned} \quad (29)$$

For the term associated with the maximum interference over the unit sphere surface, we have from (28) that

$$\frac{(\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}} (MAI' + SCI) \leq \log(n) (c_{15} + c_{16}). \quad (30)$$

Thus, from (29) and (30), and by taking $B \geq \frac{c_{18} n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}}$, for some positive constant c_{18} and n sufficiently large, we obtain

$$\begin{aligned} \frac{(\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}} (MAI' + SCI) &\leq \log(n) (c_{15} + c_{16}) \\ &\leq \frac{BN_o (\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}}. \end{aligned} \quad (31)$$

The term $\frac{BN_o (\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}}$ becomes dominant in the denominator of (29) when $B \geq \frac{c_{18} n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}}$ and $n \rightarrow \infty$. Consequently, for $B \geq \frac{c_{18} n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}}$, the link's Shannon capacity as $n \rightarrow \infty$ is given by (32) in which c_{19} is a positive constant.

Eq. (32) is the link's Shannon capacity obtained from the noise dominance over interference due to large bandwidth expansion. Note that any increase in B beyond $\frac{c_{18} n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}}$ does not change the value of this capacity.

The bandwidth expansion associated to this capacity, in which $B \geq \frac{c_{18} n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}}$, is given by

$$\Delta W = BM = \Omega \left[\frac{n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}-1}} \right]. \quad (33)$$

To obtain the throughput behavior, note that each cell has one node *whp*, and any node in this cell can be the head node to relay all the traffic the cell must handle, while the other nodes can simply serve as sources or destinations. Accordingly, analogous to (21), the per source-destination throughput is given *whp* by

$$\lambda(n) = \frac{R_{ij} \delta \mathbb{P}\{Z \geq 1\}}{\# \text{ of served routes}}, \quad (34)$$

where $\mathbb{P}\{Z \geq 1\} \xrightarrow{n \rightarrow \infty} 1$, and δ is a constant that depends on c_{12} and can be computed based on the channel access scheme employed [1].

The number of routes served by any cell is a consequence of the routing strategy. As mentioned before, the routing of packets is done through multiple hops along cells following the minimum distance path from source to destination, i.e., every packet follows the straight line segment connecting the source to its destination. Therefore, the traffic to be carried by any cell is proportional to the number of straight line segments passing through the cell. Accordingly, the number of routes intersecting any cell is bounded by the following lemma, which was proved by Gupta and Kumar [1].

Lemma 2 *The total number of source-destination lines (i.e., routes) intersecting every cell in the random network can be bounded whp by*

$$\sup_{V \in \mathcal{V}_n} (\text{Number of routes intersecting } V) \leq c_{20} \sqrt{n \log(n)}. \quad (35)$$

Therefore, using the network model assumptions provided in this Section, from (32) and (34), and from Lemma 2, we proved the following Theorem.

Theorem 2 *The static random wireless ad hoc network using CDMA and SIC attains whp the following per source-destination throughput lower-bound*

$$\lambda(n) = \Omega \left[\frac{n^{\frac{\alpha-1}{2}}}{(\log(n))^{\frac{\alpha+1}{2}}} \right]. \quad (36)$$

Theorem 2 provides the same throughput lower-bound order as that obtained by Negi and Rajeswaran [12], which corroborates the capacity analysis technique employed throughout this paper. However, our bandwidth expansion associated to this lower-bound, given by (33), is much smaller than the $\Theta(n^2 \log(n)^{\frac{\alpha}{2}})$ required by Negi and Rajeswaran [12] because we take advantage of SIC. SIC allows every node in the network to successfully receive the packets from its close neighbors increasing the minimum distance of the destructive interferers. In our case, the closest destructive interferer is located *whp* at distance $\Omega(\rho(n)) = \Omega(\sqrt{\log(n)/n})$ due to SIC, while in [12] this distance is $\Omega(1/n\sqrt{\log(n)})$ *whp* (see Lemma 1 in [12]).

The strategies used to obtain the throughput lower-bound here and in [12] assume that each node transmits at constant power and its packets follow the minimum distance path to the destination. However, it was also shown by Negi and Rajeswaran [12] that, if transmission power control is allowed, then a minimum power route to destination can be obtained (not necessarily equal to the minimum distance path) which

$$\begin{aligned}
R_{ij} &= B \log \left(1 + \frac{c_{17}}{\frac{BN_o(\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}} + \frac{(\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}} (MAI' + SCT)} \right) \\
&= \frac{n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}}} \frac{B}{\frac{n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}}}} \log \left(1 + \frac{c_{17}}{\frac{BN_o(\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}} + \frac{(\log(n))^{\frac{\alpha}{2}}}{n^{\frac{\alpha}{2}}} (MAI' + SCT)} \right) = c_{19} \frac{n^{\frac{\alpha}{2}}}{(\log(n))^{\frac{\alpha}{2}}}. \quad (32)
\end{aligned}$$

$\xrightarrow{n \rightarrow \infty} c_{19}$

provides an upper-bound for the throughput. It is the objective of our future research to allow power control in our analysis and investigate the associated behavior of the throughput and bandwidth expansion considering CDMA and SIC.

B. The Mobile Network Case

A direct comparison between many-to-many communication and the strategy proposed by Grossglauer and Tse [2] is not appropriate even after applying CDMA and bandwidth expansion, because their model does not require the use of cell we assume to enable frequency reuse. Accordingly, we extend Grossglauer and Tse's network model by introducing cells in which nodes are endowed with FDMA/CDMA-SIC and GPS capabilities, such that every node behaves simultaneously like sender and receiver of data packets for each communication session. Therefore, another comparison, not necessarily based on the physical layer properties (like link's Shannon capacity or bandwidth expansion), is more suitable.

We have recently presented a cell description [4] for Grossglauer and Tse's scheme [2] using assumptions that are similar to those used by El Gamal et al. [5]. Because only one half-duplex data channel is used for the entire network in Grossglauer and Tse's model [2], a node cannot be sender and receiver simultaneously, but rather every node behaves like either a sender or a receiver for each communication session. Accordingly, $r_o = 1/\sqrt{\pi\theta n}$ determines a cell in such a model [2], [4] for a uniform distribution of the nodes, where the parameter $\theta \in (0, 1)$ is defined as the fraction of sender nodes n_S in the network. Therefore, $n_S = \theta n$, and $n_R = (1 - \theta)n$ is the fraction of receiver nodes. It has been shown [5], [4] that the per source-destination throughput is proportional to the fraction of cells in the network that can successfully forward packets. In the work by Grossglauer and Tse [2] and in our previous work [4], only the cells containing exactly one sender (i.e., $L = 1$) and at least one receiver (i.e., $K \geq 1$) are able to forward packets, because no SIC capability is assumed, and therefore, the cells containing more than one sender present transmission collisions, preventing successful relaying of packets. Similarly to what was done to obtain (5), it can be shown that, for Grossglauer and Tse's scheme [2], we have that [4]

$$\mathbb{P}\{L = 1, K \geq 1\} = \frac{1}{\theta} e^{-1/\theta} (1 - e^{-1/\theta}) \text{ as } n \rightarrow \infty. \quad (37)$$

With many-to-many communication, in order to obtain the same cell size as in [2] and [4], i.e., $a(n) = \pi r_o^2 = \frac{1}{\theta n} = \frac{1}{\phi n}$, we must set $\theta = \phi$, and use a finite bandwidth expansion. In addition, all cells containing at least two nodes are able

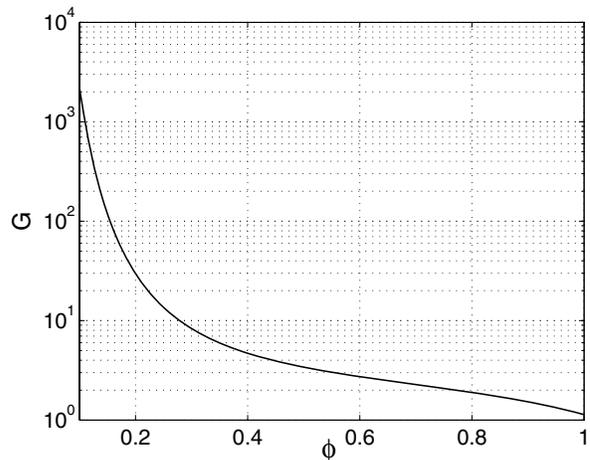


Fig. 7. Performance gain (G) for fraction of cells that successfully forward packets in *many-to-many communication* compared to Grossglauer and Tse's scheme [2].

to successfully forward packets in many-to-many communication. Thus, from (5), $\mathbb{P}\{Z \geq 2\} = (1 - e^{-1/\phi} - \frac{1}{\phi} e^{-1/\phi})$ as $n \rightarrow \infty$. Hence, our collaboration-driven strategy provides the following performance gain G over the Grossglauer and Tse's scheme [2] based on a comparison of the fraction of cells that successfully forward packets as $n \rightarrow \infty$,

$$G = \frac{\mathbb{P}\{Z \geq 2\}}{\mathbb{P}\{L=1, K \geq 1\}} = \frac{1 - e^{-1/\phi} - \frac{1}{\phi} e^{-1/\phi}}{\frac{1}{\phi} e^{-1/\phi} (1 - e^{-1/\phi})}. \quad (38)$$

Fig. 7 illustrates the behavior of the gain G given in (38) as a function of ϕ . Note that $G > 1 \forall \phi \in (0, 1)$. This gain shows that the throughput is improved by a constant factor compared to the results in [2] and [4] under similar bandwidth expansion. There is additional gain in the link Shannon capacity, as a constant gain factor, due to the use of SIC and the improvement in SNIR. However, an exact computation of this constant factor turns out to be a tedious task.

VI. CONCLUSIONS

It was shown that the Shannon capacity and per source-destination throughput can increase in wireless ad hoc networks by employing mobility, FDMA/CDMA, SIC, and one-time relaying of packets taking advantage of many-to-many communication among nodes. Such performance is attained by using successive interference cancellation and distinct codes among close neighbors, which is enabled by running a simple neighbor-discovery protocol. Accordingly, interference from

close neighbors is no longer harmful, but rather endowed with valuable data. Many-to-many communication employs multi-packet transmissions (MPTs) and multi-packet receptions (MPRs) which can provide significant improvement in scaling laws for future networks [27]. Also, because multi-copy relaying of packets is employed, the delay performance is improved and follows the description given in [4]. The overall improvement in the network performance is obtained at a cost of increased processing complexity in the nodes. Furthermore, the principles of many-to-many communication are applied to static [1], [12] and mobile [2], [4] networks. It is shown that, by using this approach, similar capacity of [12] can be attained with much smaller bandwidth expansion. We have shown that many-to-many communication improves the throughput of mobile wireless networks by a constant factor compared to the results in [2] and [4] under similar bandwidth expansion.

ACKNOWLEDGMENTS

This work was supported in part by CAPES and CNPq/Brazil, by the U.S. Army Research Office under grants W911NF-04-1-0224 and W911NF-05-1-0246, by the National Science Foundation under grant CCF-0729230, by the Defense Advanced Research Projects Agency through Air Force Research Laboratory Contract FA8750-07-C-0169, and by the Baskin Chair of Computer Engineering.

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Renato M. de Moraes received the Bachelor degree in electrical engineering from Federal University of Pernambuco-Brazil in 1996, and the Master degree from UNICAMP-Sao Paulo-Brazil in 1998. He obtained his PhD degree in EE from University of California, Santa Cruz, in 2005. Currently, Renato is an Associate Professor at the Department of Computing and Systems at the University of Pernambuco, Brazil. His areas of interest include Ad hoc Networks, Information Theory, and Communication Networks.



Hamid R. Sadjadpour received his B.S. and M.S. degrees from Sharif University of Technology with high honor and Ph.D. degree from University of Southern California in 1986, 1988, and 1996 respectively. After graduation, he joined AT&T Shannon Lab. in Florham Park, NJ as a member of technical staff, later senior technical staff member, and finally Principal member of technical staff until 2001. In fall 2001, he joined University of California, Santa Cruz (UCSC) where he is now an Associate professor. He has served as technical program committee member in numerous conferences. He has been also Guest editor of EURASIP on special issue on Multicarrier Communications and Signal Processing in 2003 and special issue on Mobile Ad Hoc Networks in 2006, and is currently Associate editor for Journal of Communications and Networks (JCN). He has published more than 110 publications. His research interests include scaling laws for wireless ad hoc networks, performance analysis of ad hoc and sensor networks, MAC and routing protocols for MANETs, and space-time signal processing. He is the co-recipient of International Symposium on Performance Evaluation of Computer and Telecommunication Systems (SPECTS) 2007 best paper award and IEEE Fred W. Ellersick Award for Best Unclassified Paper at the 2008 Milcom conference. He holds more than 13 patents, one of them adopted in the spectrum management of T1.E1.4 standard. He holds more than 13 patents, one of them accepted in spectrum management of T1.E1.4 standard.



J. J. Garcia-Luna-Aceves holds the Jack Baskin Chair of Computer Engineering at the University of California, Santa Cruz (UCSC), and is a Principal Scientist at the Palo Alto Research Center (PARC). Prior to joining UCSC in 1993, he was a Center Director at SRI International (SRI) in Menlo Park, California. He has been a Visiting Professor at Sun Laboratories and a Principal of Protocol Design at Nokia.

Dr. Garcia-Luna-Aceves has published a book, more than 370 papers, and 26 U.S. patents. He has directed 26 Ph.D. theses and 22 M.S. theses since he joined UCSC in 1993. He has been the General Chair of the ACM MobiCom 2008 Conference; the General Chair of the IEEE SECON 2005 Conference; Program Co-Chair of ACM MobiHoc 2002 and ACM MobiCom 2000; Chair of the ACM SIG Multimedia; General Chair of ACM Multimedia '93 and ACM SIGCOMM '88; and Program Chair of IEEE MULTIMEDIA '92, ACM SIGCOMM '87,

and ACM SIGCOMM '86. He has served in the IEEE Internet Technology Award Committee, the IEEE Richard W. Hamming Medal Committee, and the National Research Council Panel on Digitization and Communications Science of the Army Research Laboratory Technical Assessment Board. He has been on the editorial boards of the IEEE/ACM TRANSACTIONS ON NETWORKING, the MULTIMEDIA SYSTEMS JOURNAL, and the JOURNAL OF HIGH SPEED NETWORKS.

He is a Fellow of the IEEE and is listed in Marquis Who's Who in America and Who's Who in The World. He is the co-recipient of the IEEE Fred W. Ellersick Award for best unclassified paper at IEEE MILCOM 2008. He is also co-recipient of Best Paper awards at the IEEE MASS 2008, SPECTS 2007, IFIP Networking 2007, and IEEE MASS 2005 conferences, and of the Best Student Paper Award of the 1998 IEEE International Conference on Systems, Man, and Cybernetics. He received the SRI International Exceptional-Achievement Award in 1985 and 1989.